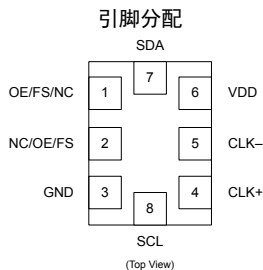
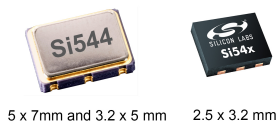


Ultra Series™ 晶体振荡器系列

Si544 数据表

超低抖动 I2C 可编程 XO (150 fs), 0.2 至 1500 MHz

Si544 Ultra Series™ 振荡器系列采用 Silicon Laboratories 先进的第四代 DSPLL® 技术，提供任意输出频率的超低抖动和低相位噪声时钟。这款设备由用户通过简单的 I2C 命令进行编程，可提供频率范围在 0.2 至 1500 MHz 之间的任意频率，分辨率小于 1 ppb，可以在整个工作范围内实现整数和小数频率的超低抖动。Si544 振荡器系列提供出色的可靠性、频率稳定性和抗老化性能。片上电源滤波可以实现行业领先的电源噪声抑制特性，简化了使用开关式电源的噪声系统生成低抖动时钟的任务。Si544 振荡器系列大幅简化供应链，使 Silicon Labs 在收到订单后 1-2 周内即可将定制频率样品送达。不同于传统的 XO，Si544 振荡器系列无需使用不同的晶体实现不同的输出频率，而使用单一晶体和基于 DSPLL IC 的方法提供所需输出频率。Si544 振荡器系列经工厂配置，可以满足各种各样的用户规格，包括启动频率、I2C 地址、输出格式和 OE 引脚位置/极性。特殊配置在发货时经过出厂前编程，消除了与定制频率振荡器有关的长交付周期。



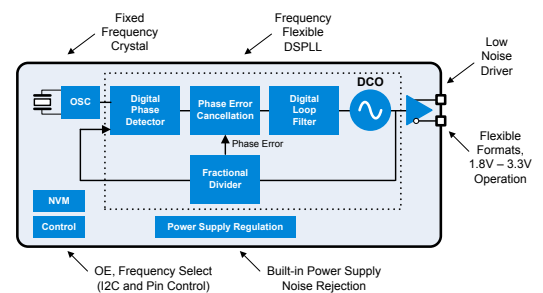
主要特点

- I2C 可编程为 0.2 至 1500 MHz 频率范围内的任何频率，分辨率小于 1 ppb
- 极低抖动：150 fs RMS (典型值，12 kHz - 20 MHz)
- 最多可配置 4 个引脚可选启动频率
- I2C 接口支持 100 kbps、400 kbps 和 1 Mbps (超快速模式)
- 出色的 PSNR 和电源噪声抗扰度：-80 dBc (典型值)
- 相同部件编号可实现 3.3 V、2.5 V 和 1.8 V 的供电电压电源操作
- 提供 LVPECL、LVDS、CML、HCSL、CMOS 和双路 CMOS 输出选项
- 2.5x3.2、3.2x5、5x7 mm 封装选项
- 样品交付周期为 1-2 周

应用

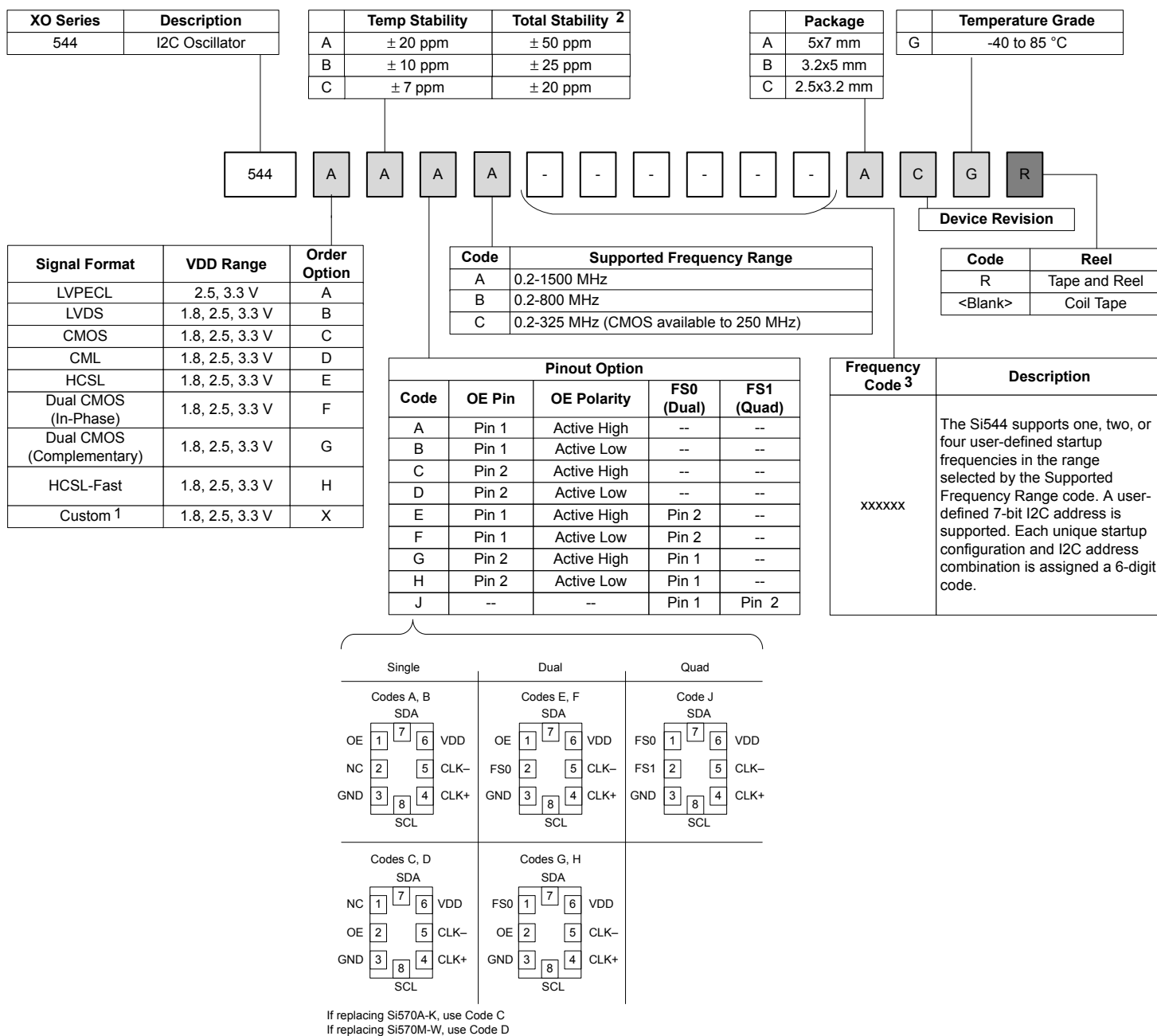
- 100G/200G/400G 相干光学 OTN, PAM4
- 10G/40G/100G 光以太网
- 3G-SDI/12G-SDI/24G-SDI 广播视频
- 服务器、开关、存储、搜索加速
- 测试和测量
- FPGA/ASIC 时钟设计

引脚编号	说明
1, 2	订购可选项 OE 表示“输出使能”；FS 表示“频率选择”；NC 表示“无连接”
3	GND 表示“接地”
4	CLK+ 表示“时钟输出”
5	CLK- 表示“互补时钟输出”。CMOS 输出格式下不可用。
6	VDD 表示“电源电压”
7	SDA = I2C 串行数据
8	SCL = I2C 串行时钟



1. Ordering Guide

The Si544 XO supports a variety of options including startup frequency, output format, and OE pin location/polarity, as shown in the chart below. Specific device configurations are programmed into the part at time of shipment, and samples are available in 1-2 weeks. Silicon Laboratories provides an online part number configuration utility to simplify this process. Refer to www.silabs.com/oscillators to access this tool and for further ordering instructions.



Notes:

- Contact Silicon Labs for non-standard configurations.
- Total stability includes temp stability, initial accuracy, load pulling, VDD variation, and 20 year aging at 70 °C.
- Create custom part numbers at www.silabs.com/oscillators.

1.1 Technical Support

Frequently Asked Questions (FAQ)	www.silabs.com/Si544-FAQ
Oscillator Phase Noise Lookup Utility	www.silabs.com/oscillator-phase-noise-lookup
Quality and Reliability	www.silabs.com/quality
Development Kits	www.silabs.com/oscillator-tools

2. Electrical Specifications

Table 2.1. Electrical Specifications

 $V_{DD} = 1.8\text{ V}, 2.5\text{ or }3.3\text{ V} \pm 5\%$, $T_A = -40\text{ to }85\text{ }^\circ\text{C}$

Parameter	Symbol	Test Condition/Comment	Min	Typ	Max	Unit
Temperature Range	T_A		-40	—	85	$^\circ\text{C}$
Frequency Range	F_{CLK}	LVPECL, LVDS, CML	0.2	—	1500	MHz
		HCSL	0.2	—	400	MHz
		CMOS, Dual CMOS	0.2	—	250	MHz
Supply Voltage	V_{DD}	3.3 V	3.135	3.3	3.465	V
		2.5 V	2.375	2.5	2.625	V
		1.8 V	1.71	1.8	1.89	V
Supply Current	I_{DD}	LVPECL (output enabled)	—	100	145	mA
		LVDS/CML (output enabled)	—	75	111	mA
		HCSL (output enabled)	—	80	125	mA
		HCSL-Fast (output enabled)	—	88	137	mA
		CMOS (output enabled)	—	74	108	mA
		Dual CMOS (output enabled)	—	80	125	mA
		Tristate Hi-Z (output disabled)	—	64	100	mA
Temperature Stability		Frequency stability Grade A	-20	—	20	ppm
		Frequency stability Grade B	-10	—	10	ppm
		Frequency stability Grade C	-7	—	7	ppm
Total Stability ¹	F_{STAB}	Frequency stability Grade A	-50	—	50	ppm
		Frequency stability Grade B	-25	—	25	ppm
		Frequency stability Grade C	-20	—	20	ppm
Rise/Fall Time (20% to 80% V_{PP})	T_R/T_F	LVPECL/LVDS/CML	—	—	350	ps
		CMOS / Dual CMOS ($C_L = 5\text{ pF}$)	—	0.5	1.5	ns
		HCSL, $F_{CLK} > 50\text{ MHz}$	—	—	550	ps
		HCSL-Fast, $F_{CLK} > 50\text{ MHz}$	—	—	275	ps
Duty Cycle	D_C	All formats	45	—	55	%
Output Enable (OE), Frequency Select (FS0, FS1) ²	V_{IH}		$0.7 \times V_{DD}$	—	—	V
	V_{IL}		—	—	$0.3 \times V_{DD}$	V
	T_D	Output Disable Time, $F_{CLK} > 10\text{ MHz}$	—	—	3	μs
	T_E	Output Enable Time, $F_{CLK} > 10\text{ MHz}$	—	—	20	μs
	T_{FS}	Settling Time after FS Change	—	—	10	ms
Powerup Time	t_{OSC}	Time from $0.9 \times V_{DD}$ until output frequency (F_{CLK}) within spec	—	—	10	ms

Parameter	Symbol	Test Condition/Comment	Min	Typ	Max	Unit
Powerup VDD Ramp Rate	V_{RAMP}	Fastest VDD ramp rate allowed on startup	—	—	100	V/ms
LVPECL Output Option ³	V_{OC}	Mid-level	$V_{DD} - 1.42$	—	$V_{DD} - 1.25$	V
	V_O	Swing (diff)	1.1	—	1.9	V_{PP}
LVDS Output Option ⁴	V_{OC}	Mid-level (2.5 V, 3.3 V VDD)	1.125	1.20	1.275	V
		Mid-level (1.8 V VDD)	0.8	0.9	1.0	V
	V_O	Swing ($F_{CLK} \leq 1.4$ GHz)	0.6	0.7	0.9	V_{PP}
		Swing ($F_{CLK} > 1.4$ GHz)	0.5	0.7	0.9	V_{PP}
HCSL Output Option ⁵ HCSL-Fast Output Option ⁵	V_{OH}	Output voltage high	660	750	850	mV
	V_{OL}	Output voltage low	-150	0	150	mV
	V_C	Crossing voltage	250	350	550	mV
CML Output Option (AC-Coupled)	V_O	Swing (diff)	0.6	0.8	1.0	V_{PP}
CMOS Output Option	V_{OH}	$I_{OH} = 8/6/4$ mA for 3.3/2.5/1.8V VDD	$0.85 \times V_{DD}$	—	—	V
	V_{OL}	$I_{OL} = 8/6/4$ mA for 3.3/2.5/1.8V VDD	—	—	$0.15 \times V_{DD}$	V

Notes:

- Total Stability includes temperature stability, initial accuracy, load pulling, VDD variation, and aging for 20 yrs at 70 °C.
- OE includes a 50 k Ω pull-up to VDD for OE active high. Includes a 50 k Ω pull-down to GND for OE active low. FS0 and FS1 pins each include a 50 k Ω pull-up to VDD. NC (No Connect) pins include a 50 k Ω pull-down to GND.
- 50 Ω to $V_{DD} - 2.0$ V. Additional DC current from the output driver will flow through the 50 Ω resistors, resulting in a shift in common mode voltage. The measurements in this table have accounted for this.
- $R_{term} = 100$ Ω (differential).
- 50 Ω to GND.

Table 2.2. I2C Characteristics

$V_{DD} = 1.8, 2.5, \text{ or } 3.3 \text{ V} \pm 5\%$, $T_A = -40 \text{ to } 85$ °C

Parameter	Symbol	Test Condition/Comment	Min	Typ	Max	Unit
SDA, SCL Input Voltage High	V_{IH}		$0.70 \times V_{DD}$	—	—	V
SDA, SCL Input Voltage Low	V_{IL}		—	—	$0.30 \times V_{DD}$	V
Frequency Reprogramming Resolution	M_{RES}		—	0.026	—	ppb
Frequency Range for Small Frequency Change (Continuous Glitchless Output)		From center frequency	-950	—	+950	ppm
Settling Time for Small Frequency Change		< ± 950 ppm from center frequency	—	—	100	μ s
Settling Time for Large Frequency Change (Output Squelched during Frequency Transition). Includes 30ms for internal VCO calibration to new frequency (FCAL). See Table 5.6.		> ± 950 ppm from center frequency	—	—	40	ms

Note: For best performance, place the Si544 on a dedicated I2C bus, or isolate it via an I2C bus multiplexer to minimize noise.

Table 2.3. Clock Output Phase Jitter and PSNR $V_{DD} = 1.8, 2.5, \text{ or } 3.3 \text{ V} \pm 5\%$, $T_A = -40 \text{ to } 85 \text{ }^\circ\text{C}$

Parameter	Symbol	Test Condition/Comment	Min	Typ	Max	Unit
Phase Jitter (RMS, 12kHz - 20MHz) ¹ $F_{CLK} \geq 100 \text{ MHz}$	ϕ_J	LVPECL, HCSSL, CML	—	150	200	fs
		LVDS	—	150	220	fs
		CMOS, Dual CMOS	—	200	—	fs
Spurs Induced by External Power Supply Noise, 50 mVpp Ripple. LVDS 156.25 MHz Output	PSNR	100 kHz sine wave	—	-83	—	dBc
		200 kHz sine wave	—	-83	—	
		500 kHz sine wave	—	-82	—	
		1 MHz sine wave	—	-85	—	

Note:
1. Guaranteed by characterization. Jitter inclusive of any spurs, I2C lines not active.

Table 2.4. 3.2x5 Clock Output Phase Noise (Typical, 50ppm Total Stability Option)

Offset Frequency (f)	156.25 MHz LVDS	200 MHz LVDS	644.53125 MHz LVDS	Unit
100 Hz	-108	-106	-97	dBc/Hz
1 kHz	-124	-123	-111	
10 kHz	-132	-130	-120	
100 kHz	-137	-135	-125	
1 MHz	-146	-144	-133	
10 MHz	-160	-161	-154	
20 MHz	-161	-162	-156	
Offset Frequency (f)	156.25 MHz LVPECL	200 MHz LVPECL	644.53125 MHz LVPECL	Unit
100 Hz	-105	-105	-95	dBc/Hz
1 kHz	-122	-122	-113	
10 kHz	-132	-131	-120	
100 kHz	-136	-135	-125	
1 MHz	-145	-144	-134	
10 MHz	-160	-163	-155	
20 MHz	-161	-164	-156	

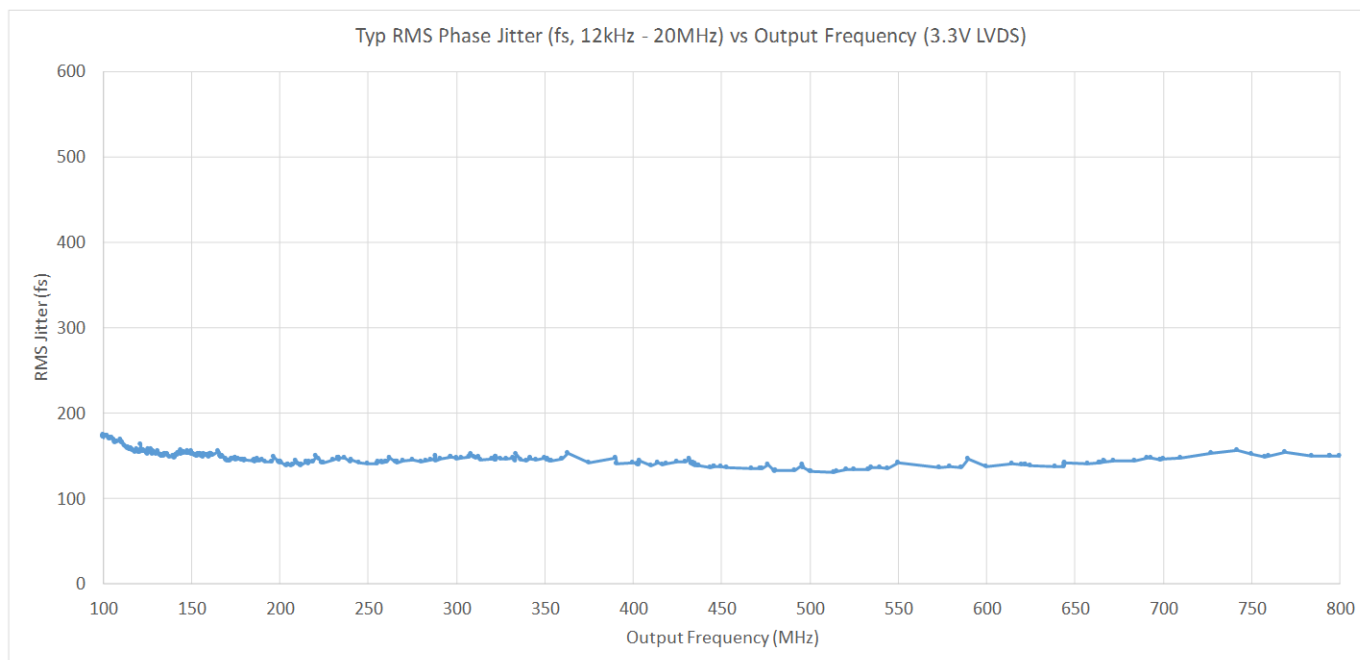


Figure 2.1. Phase Jitter vs. Output Frequency

Phase jitter measured with Agilent E5052 using a differential-to-single ended converter (balun or buffer). Measurements collected for >700 commonly used frequencies. Phase noise plots for specific frequencies are available using our free, online Oscillator Phase Noise Lookup Tool at www.silabs.com/oscillators.

Table 2.5. Environmental Compliance and Package Information

Parameter	Test Condition
Mechanical Shock	MIL-STD-883, Method 2002
Mechanical Vibration	MIL-STD-883, Method 2007
Solderability	MIL-STD-883, Method 2003
Gross and Fine Leak	MIL-STD-883, Method 1014
Resistance to Solder Heat	MIL-STD-883, Method 2036
Moisture Sensitivity Level (MSL): 3.2 x 5, 5 x 7 packages	1
Moisture Sensitivity Level (MSL): 2.5 x 3.2 package	2
Contact Pads: 3.2x5, 5x7 packages	Au/Ni (0.3 - 1.0 μm / 1.27 - 8.89 μm)
Contact Pads: 2.5x3.2 packages	Au/Pd/Ni (0.03 - 0.12 μm / 0.1 - 0.2 μm / 3.0 - 8.0 μm)
Note:	
1. For additional product information not listed in the data sheet (e.g. RoHS Certifications, MDDS data, qualification data, REACH Declarations, ECCN codes, etc.), refer to our "Corporate Request For Information" portal found here: www.silabs.com/support/quality/Pages/RoHSInformation.aspx .	

Table 2.6. Thermal Conditions

Max Junction Temperature = 125 °C

Package	Parameter	Symbol	Test Condition	Value	Unit
2.5 x 3.2 mm 8-pin DFN	Thermal Resistance Junction to Ambient	Θ_{JA}	Still Air, 85 °C	80	°C/W
	Thermal Parameter Junction to Board	Ψ_{JB}	Still Air, 85 °C	39	°C/W
	Thermal Parameter Junction to Top Center	Ψ_{JT}	Still Air, 85 °C	17	°C/W
3.2 x 5 mm 8-pin CLCC	Thermal Resistance Junction to Ambient	Θ_{JA}	Still Air, 85 °C	55	°C/W
	Thermal Parameter Junction to Board	Ψ_{JB}	Still Air, 85 °C	20	°C/W
	Thermal Parameter Junction to Top Center	Ψ_{JT}	Still Air, 85 °C	20	°C/W
5 x 7 mm 8-pin CLCC	Thermal Resistance Junction to Ambient	Θ_{JA}	Still Air, 85 °C	53	°C/W
	Thermal Parameter Junction to Board	Ψ_{JB}	Still Air, 85 °C	26	°C/W
	Thermal Parameter Junction to Top Center	Ψ_{JT}	Still Air, 85 °C	26	°C/W
Note:					
1. Based on PCB Dimensions: 4.5" x 7", PCB Thickness: 1.6 mm, Number of Cu Layers: 4.					

3. Dual CMOS Buffer

Dual CMOS output format ordering options support either complementary or in-phase signals for two identical frequency outputs. This feature enables replacement of multiple XOs with a single Si544 device.

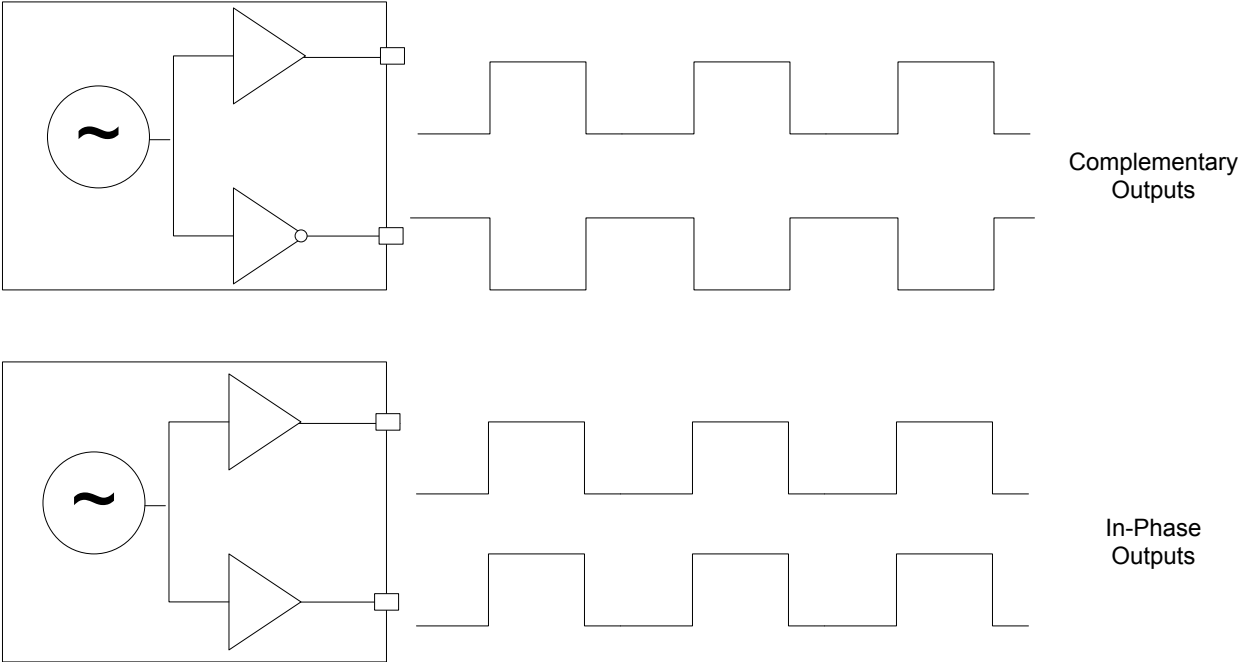


Figure 3.1. Integrated 1:2 CMOS Buffer Supports Complementary or In-Phase Outputs

4. Recommended Output Terminations

The output drivers support both AC-coupled and DC-coupled terminations as shown in figures below.

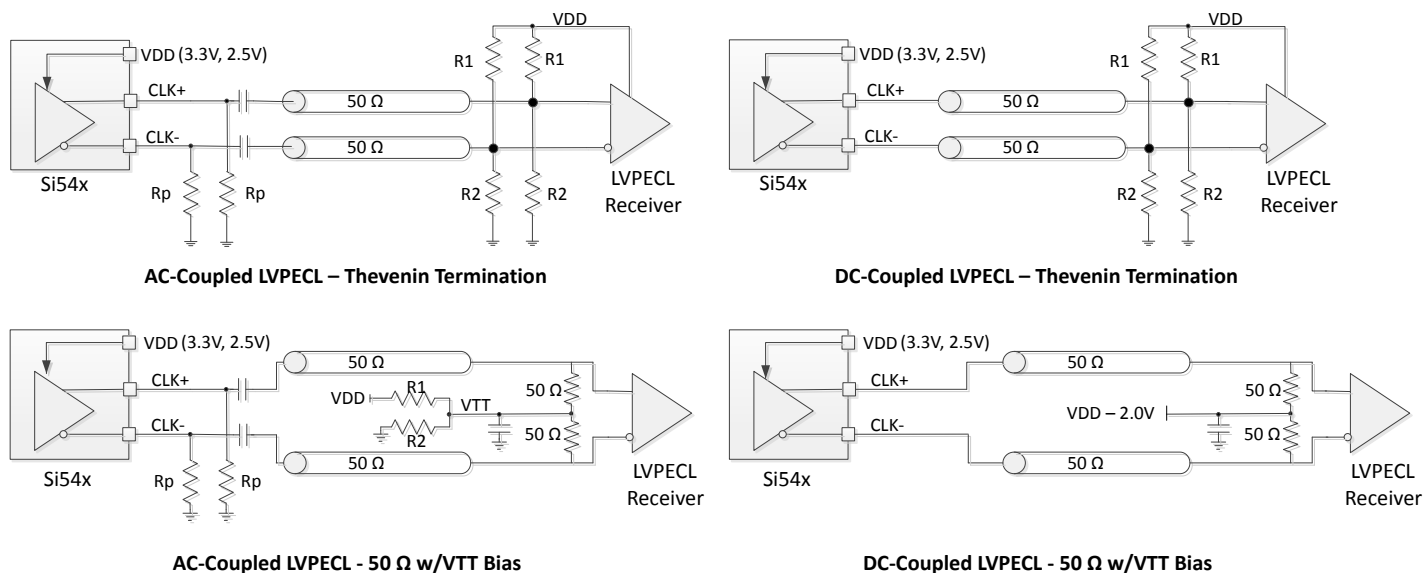


Figure 4.1. LVPECL Output Terminations

AC Coupled LVPECL Termination Resistor Values				DC Coupled LVPECL Termination Resistor Values		
VDD	R1	R2	Rp	VDD	R1	R2
3.3 V	82.5 Ω	127 Ω	130 Ω	3.3 V	127 Ω	82.5 Ω
2.5 V	62.5 Ω	250 Ω	90 Ω	2.5 V	250 Ω	62.5 Ω

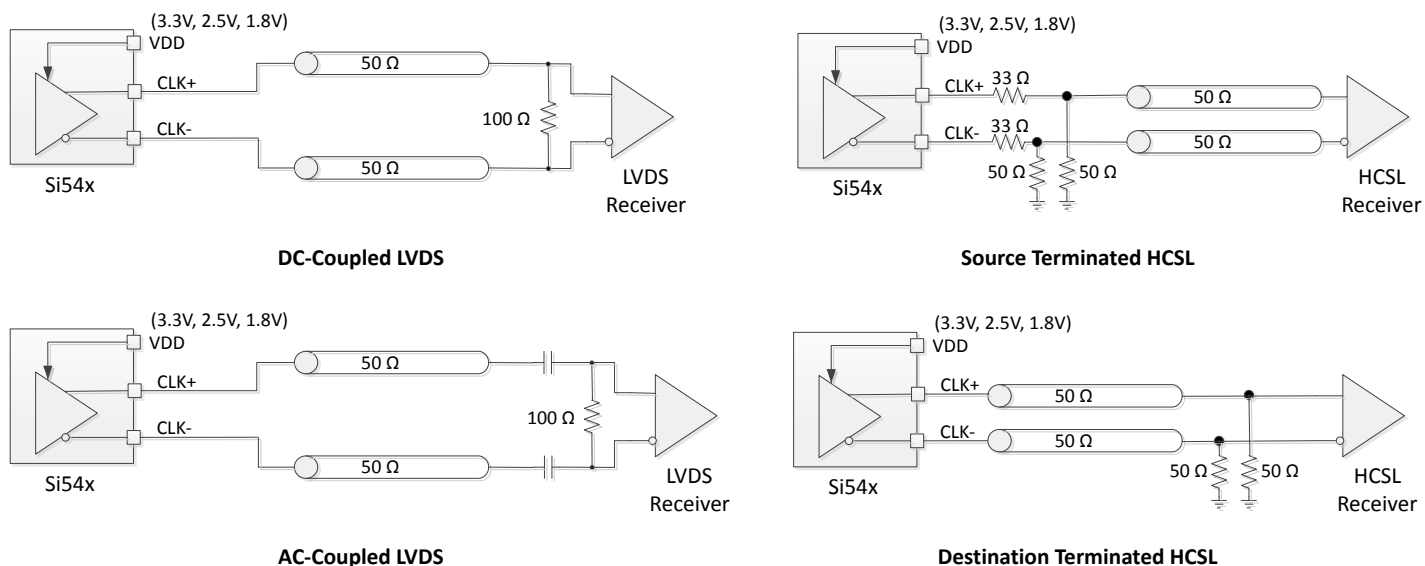


Figure 4.2. LVDS and HCSL Output Terminations

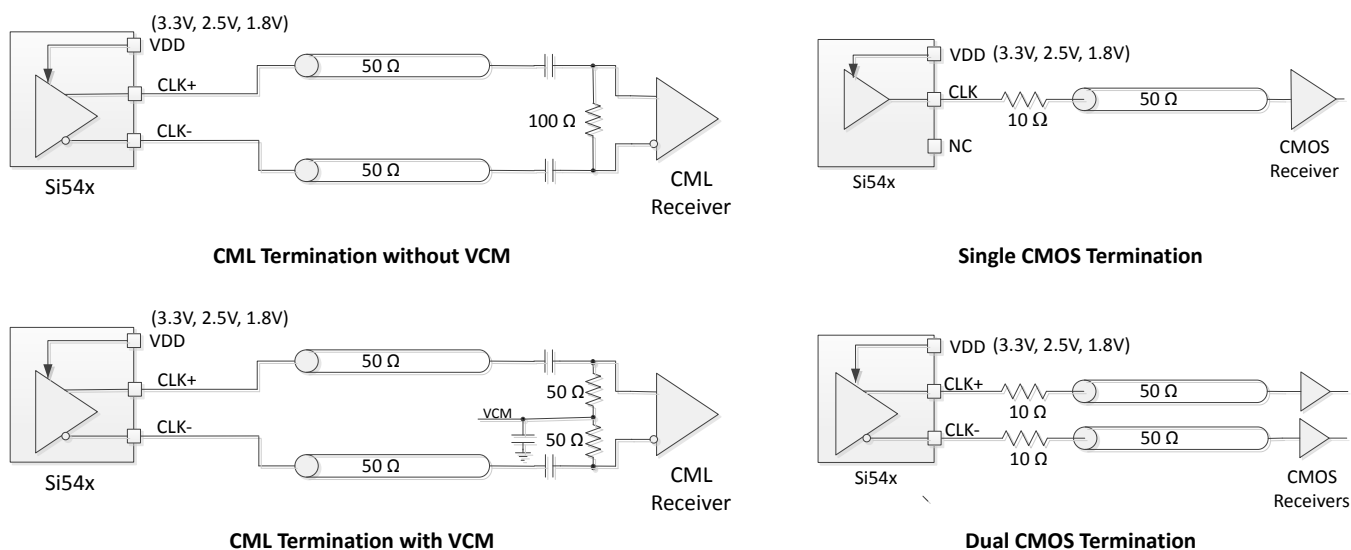


Figure 4.3. CML and CMOS Output Terminations

5. Configuring Si544 Output Frequency via I2C

The Si544 oscillator device contains a fixed frequency crystal and frequency synthesis IC using Silicon Labs patented DSPLL™ technology, all enclosed in a standard hermetically sealed crystal oscillator (XO) package. The internal crystal provides the reference frequency used by the DSPLL frequency synthesis IC. The output frequency of the Si544 oscillator device can be dynamically set via I2C register settings in the DSPLL frequency synthesis IC. DSPLL technology provides unmatched frequency flexibility with superior output jitter/phase noise performance and part per trillion frequency accuracy. This document describes how to calculate the required Si544 register values used to set device output frequency, and how to load these values into the Si544 device.

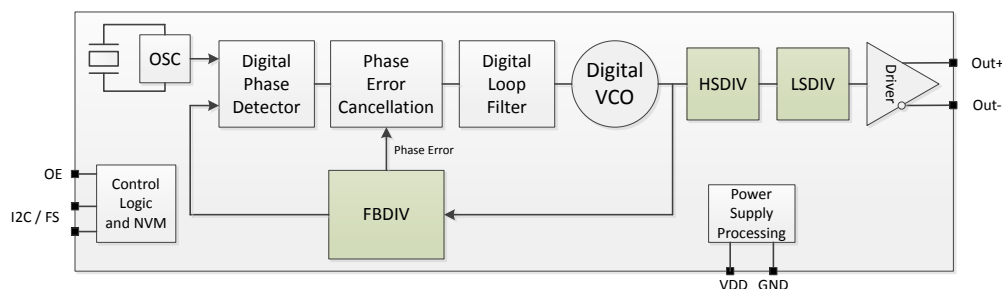


Figure 5.1. Si544 Block Diagram

The figure above is a simplified high-level block diagram of the Si544 oscillator device. The output frequency is set by a combination of three divider blocks highlighted in the above block diagram.

1. FBDIV - DSPLL™ Feedback Divider used to set Digital VCO frequency
2. HSDIV - High-Speed Output Divider
3. LSDIV - Low-Speed Output Divider

The final device output frequency is based on the digital VCO frequency divided by the product of HSDIV and LSDIV divider settings. The limits of each of these internal blocks (both digital VCO and dividers) determines the valid operating frequency range of the device.

The FBDIV divider, is a fractional fixed-point divider with a total length of 43 bits consisting of an 11-bit integer field (FBINT) and a 32 bit fractional field (FBFRAC) where total FBDIV = [FBINT].[FBFRAC] with an implied decimal point as shown. This bit format is known as an 11.32 fixed point format where the integer portion is 11 bits and fractional portion is 32 bits, for a total of 43 bits.

The HSDIV divider is an integer divider, 11 bits in length, containing a binary divider value. One noteworthy feature of the HSDIV divider is a special duty cycle correction circuit that allows **odd** divide ratios of lower divider values (5-33 only) with 50% duty cycle output. This feature is useful when LSDIV divide ratio is set to 1.

The LSDIV divider performs power-of-2 divides ranging from divide by 1 (20) to divide by 32 (25). The register controlling the LSDIV divider is 3 bits in length, holding the power-of-2 divide ratio (divider exponent). For example, if LSDIV register = 3 the LSDIV divide ratio is $2^3 = 8$. Values greater than 5 (i.e. LSDIV register = 6 or 7) still map into a divide by 32.

The tables below summarize the divider limits for LSDIV, HSDIV, FBDIV. These limits and restrictions **must** be observed when deriving divider register values, as will be explained in later sections.

Table 5.1. Si544 Divider Range Limits

Divider	Upper Limit	Lower Limit
HSDIV[10:0] (unsigned)	2046	5
LSDIV[2:0] ¹ (unsigned)	32 (2^5)	1 (2^0)
FBDIV[42:0] hex (unsigned)	7FDFFFFFFF	03C0000000
FBDIV[42:0] int.frac (unsigned)	2045.99999999976	60.0

Note:

1. LSDIV is power of 2 divider. See LSDIV table below for actual divide ratio based on LSDIV register value.

Table 5.2. Additional LSDIV and HSDIV Divider Restrictions

LSDIV Register Value	Divide Ratio	HSDIV Value Restrictions
0	1	5-33 even or odd values ¹ , 34-2046 even values only
1	2	
2	4	
3	8	
4	16	
5	32	
6	32	
7	32	
Note:		
1. HSDIV can implement low value (5-33) odd divide ratios while providing a 50% duty cycle output due to special duty cycle correction circuit.		

Note that all divider values (FBDIV, HSDIV, LSDIV) are **unsigned** and contain only positive values.

The Si544 high-performance oscillator family has three different speed grade offerings, each covering a specific frequency range. The table below outlines the output frequency range coverage by each speed grade, the corresponding min and max VCO frequency for that speed grade, and the nominal crystal frequency. The information in the table below is needed when calculating divider settings for a given device, speed grade, and output frequency.

Table 5.3. Si544 Speed Grades, Crystal Frequency, and VCO Range Limits

Device	Speed Grade	Xtal freq (MHz)	Min Output Freq (MHz)	Max Output Freq (MHz)	Min Fvco (GHz)	Max Fvco (GHz)
Si544	A	55.05	0.2	1500	10.8	12.550082103
	B	55.05	0.2	800	10.8	12.109728345
	C	55.05	0.2	325	10.8	12.109728345

5.1 Output Frequency Equations

The basic equations used to derive the output frequency are given below and can be easily inferred from the device block diagram in [Figure 5.2 Si544 Frequency Definition Block Diagram on page 14](#). Equation 1 is the relationship between the output frequency (F_{out}), and the VCO frequency (F_{vco}) and total output divider ratio ($HSDIV * LSDIV$). Equation 2 is the relationship between the VCO frequency (F_{vco}), the fixed crystal oscillator frequency (F_{osc}), and the feedback divider ($FBDIV$).

$$F_{out} = F_{vco} / (HSDIV \times LSDIV)$$

Equation 1

$$F_{vco} = (F_{osc} \times FBDIV)$$

Equation 2

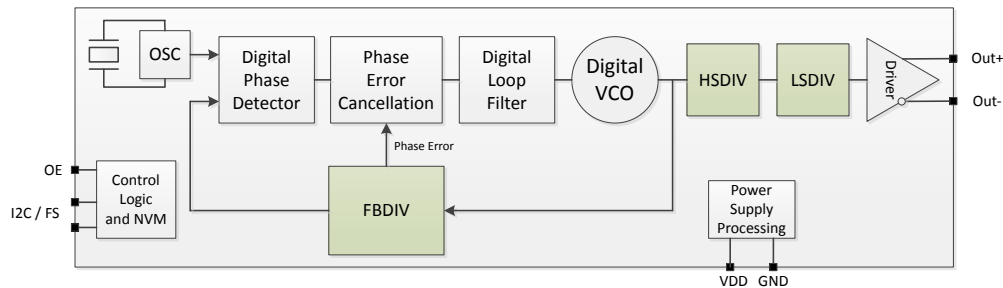


Figure 5.2. Si544 Frequency Definition Block Diagram

Equation 3a is a rearranged Equation 1 to solve for the total output divider ($HSDIV * LSDIV$) given F_{out} and F_{vco} . Equation 3b is rearranged again solving for F_{vco} given F_{out} and ($HSDIV * LSDIV$).

$$(HSDIV \times LSDIV) = F_{vco} / F_{out}$$

Equation 3a

$$F_{vco} = F_{out} \times (HSDIV \times LSDIV)$$

Equation 3b

Equation 4 is a rearranged Equation 2 to now solve for $FBDIV$ given F_{vco} and F_{osc} .

$$FBDIV = F_{vco} / F_{osc}$$

Equation 4

Equations 3a, 3b, and 4 will be used in the process of deriving the required divider values to provide a desired output frequency. The basic process is outlined below.

5.2 General Process Steps for Divider Calculation

1. Estimate a theoretical total output divider value ($HSDIV * LSDIV$) based on desired F_{out} while targeting the minimum valid F_{vco} frequency using Eqn. 3a and [Table 5.3 Si544 Speed Grades, Crystal Frequency, and VCO Range Limits on page 13](#). Use floating point calculations for this step.
 - Result: Floating point value of total ($HSDIV * LSDIV$).
2. Derive a valid LSDIV divider value based on LSDIV and HSDIV divider limitations **using the lowest possible value for LSDIV**. For example, if $(HSDIV * LSDIV) = 8.22$, use $LSDIV = 1$ and $HSDIV = 8.22$ versus $LSDIV = 2$ and $HSDIV = 4.11$.
 - Result: Valid LSDIV value.
3. Using LSDIV value from #2 above, find nearest valid **integer** HSDIV divider value resulting in F_{vco} being **equal to or greater than $F_{vco\ min}$** , which observing all HSDIV limitations. Use Eqns. 3a/3b as necessary.
 - Result: Valid HSDIV value.
4. With valid integer HSDIV and LSDIV values, calculate the required F_{vco} frequency with Eqn. 3b. (F_{vco} must remain in valid range per [Table 5.3 Si544 Speed Grades, Crystal Frequency, and VCO Range Limits on page 13](#).)
 - Result: Valid VCO frequency.
5. With the derived valid F_{vco} frequency, use Eqn 4 to calculate required FBDIV based on device specific F_{osc} frequency from [Table 5.3 Si544 Speed Grades, Crystal Frequency, and VCO Range Limits on page 13](#).
 - Result: Valid FBDIV value
6. At this point all FBDIV, HSDIV and LSDIV values required to generate the desired output frequency have been calculated. These three divider values must be now be appropriately formatted to fit the register format expected by the device. This is described in a later section.

5.3 Example: Deriving Si544 Divider Settings for 156.75 MHz Output

The general process of deriving divider values for a specific output frequency is outlined in the previous section and now will be used in this example. To reiterate, all calculations must be done while observing divider limits and valid VCO frequency range limits for your device. In this example, the device is Si544 and with a desired output frequency of 156.75 MHz, the speed grade required will be “C” or better. (One important note: All divider and register settings derived for any speed grade will work without modification for all **faster** speed grades on the same base part number device.)

Example VB code that implements the following divider calculation process is given in 5.8 Si544 Frequency Planner VB Code and can be used for implementing any supported output frequency.

Step 1: Find the valid theoretical lower limit of the total output divider (HSDIV*LSDIV) based on the desired output frequency and lowest valid VCO frequency. This will bias the divider solution to the lowest possible VCO frequency since this will provide the best performance solution.

Given the valid Si544 VCO range is 10.8000 GHz to 12.1097 GHz, the minimum theoretical values for (HSDIV * LSDIV) for the example 156.75 MHz output frequency are given in Equation 3:

$$\text{Minimum (HSDIV*LSDIV)} = (10.8000 \text{ GHz} / 156.75 \text{ MHz}) = \underline{\mathbf{68.89952\dots}}$$

Step 2: Find valid LSDIV divisor value given minimum (HSDIV*LSDIV) from step 1. For best performance, preference should be given to implementation of the total output divider (HSDIV*LSDIV) using HSDIV with LSDIV divide ratio = 1, if possible. Use LSDIV divide ratios > 1 only if HSDIV alone cannot implement the required output divider. Since the total (HSDIV*LSDIV) value of 68.8995... is less than the HSDIV maximum divider value of 2046, the LSDIV divide **ratio** value will be 1, which corresponds to a LSDIV register setting of 0, since the LSDIV divider can only be a power of 2 value (see [Table 5.2 Additional LSDIV and HSDIV Divider Restrictions on page 13](#) for valid LSDIV settings).

$$\text{LSDIV divide ratio} = \underline{\mathbf{1}}, \text{ therefore } \mathbf{LSDIV \text{ register value} = 0}$$

Step 3: Find HSDIV divisor value. Given LSDIV = 1, HSDIV must implement 68.8995... or greater. Since HSDIV is an integer divider, the next greatest integer is 69. But, checking valid HSDIV values when LSDIV divide ratio = 1, we see 69 is NOT valid since it is greater than 33 and an odd value. This means the next greater integer value must be used, which is 70 (now even value). Note that 68 would **not** be valid since 68 is less than 68.8995... and would result in a VCO frequency below the lower VCO frequency limit.

$$\text{HSDIV divide ratio} = \underline{\mathbf{70}}, \text{ which gives } \mathbf{HSDIV \text{ register value} = 70 \text{ decimal (or hex value} = \mathbf{0x46)}$$

Step 4: Calculate a valid VCO frequency and corresponding floating point FBDIV value. Given the calculated output divider value (HSDIV*LSDIV) = 70, the VCO frequency must be set to (156.75 MHz * 70) = **10.9725 GHz**. Note that 10.9725 GHz is indeed within the valid VCO frequency range per [Table 5.3 Si544 Speed Grades, Crystal Frequency, and VCO Range Limits on page 13](#).

$$F_{\text{vco}} = \mathbf{10.9725 \text{ GHz}}$$

Step 5: Calculate the FBDIV value necessary to provide a 10.9725 GHz Fvco using a 55.05 MHz crystal as reference (Si544 device). The floating point FBDIV value required to attain 10.9725 GHz with a 55.05 MHz crystal reference can be calculated as follows:

$$\mathbf{FBDIV \text{ (float)}} = 10.9725 \text{ GHz} / 55.05 \text{ MHz} = \underline{\mathbf{199.318801089918}}$$

Step 6: Format each divider value into the required register format. LSDIV and HSDIV are simply binary values and can be directly used. FBDIV must first be put into 11.32 fixed point format. Converting the floating point FBDIV value into the 11.32 fixed point hex value required by the Si544 is done as follows:

Integer value = **199** decimal. Convert **199** to 11 bit hex = **0x0C7**. This is FBINT.

Fractional value = **0.318801089918**. Multiply fractional value by $2^{32} = 1369240255.12805$. Now extract only the **integer** part of the result which is 1369240255. Convert 1369240255 to 32 bit hex = **0x519CF2BF**. This is FBFRAC.

The resulting 11.32 fixed point hex number is therefore:

$$\mathbf{FBDIV = FBINT.FBFRAC = 0x0C7519CF2BF}$$

At this point we have calculated all the required divider values. The table below summarizes the resulting divider values for implementing a 156.75 MHz output clock on the Si544.

Table 5.4. Divider Register Values for Si544 Configured for 156.75 MHz Output Clock

Divider Register	Decimal Value	Hex Value	Reg Length (bits)
LSDIV	0	0x0	3
HSDIV	70	0x046	11
FBDIV	199.318801089918	0x0C7519CF2BF	43 (11+32)

5.4 Mapping Divider Settings into Register Values

For the previous 156.75 MHz example, the divider value to register mapping is shown in the table below. Note that Register 24 is a packed register and contains bits from both LSDIV and HSDIV registers as follows: LSDIV[2:0] maps into Reg24[6:4] and HSDIV[10:8] maps into Reg24[2:0]. Note that bits Reg24[7] and Reg24[3] are not used and indicated with 'x' in the RegName field below. See also the Register Map Reference section for specific bit positioning within registers.

Table 5.5. Si544 Divider Register Values for 156.75 MHz Output Clock Configuration

Register (Decimal)	Hex Value	Reg Name
23	46	HSDIV[7:0]
24	00	x:LSDIV[2:0]:x:HSDIV[10:8]
26	BF	FBDIV[7:0]
27	F2	FBDIV[15:8]
28	9C	FBDIV[23:16]
29	51	FBDIV[31:24]
30	C7	FBDIV[39:32]
31	00	FBDIV[42:40]

5.5 I2C Register Write Procedure to Set Output Frequency

After the frequency setting registers (Reg 23-Reg31) are calculated, there is a procedure that must be followed involving other specific control registers for the device to properly use the new frequency setting registers. Simply writing Reg23-Reg31 is not enough. The following procedure must be performed as shown to properly configure the Si544 for the desired output frequency. In other words, all the following register writes must be done, and **in the exact sequence shown**.

This programming sequence consists of three distinct phases.

1. Writing to specific registers to get the device ready to be updated.
2. Writing the calculated frequency (divider) settings for the desired output frequency.
3. Writing to specific registers necessary to start-up the device after divider registers have been updated. The new output frequency will appear on output.

The divider values shown in the table below are for the previously described Si544 example for an output frequency of 156.75 MHz (for other frequencies, replace the divider values in registers 23-31 with values specific to your frequency requirements).

Table 5.6. Si544 Register Write Sequence to Set Output Frequency

Register (decimal)	Write Data (hex)	Description	Purpose
255	0x00	Set page register to point to page 0	Get Device Ready for Update
69	0x00	Disable FCAL override (to allow FCAL for this Freq Update)	
17	0x00	Synchronously disable output	
23	0x46	HSDIV[7:0]	Update Dividers
24	0x00	LSDIV[2:0]:HSDIV[10:8]	
26	0xBF	FBDIV[7:0]	
27	0xF2	FBDIV[15:8]	
28	0x9C	FBDIV[23:16]	
29	0x51	FBDIV[31:24]	
30	0xC7	FBDIV[39:32]	
31	0x00	FBDIV[42:40]	
7	0x08	Start FCAL using new divider values	
—	—	Internal FCAL VCO calibration (30 ms delay)	
17	0x01	Synchronously enable output	

5.6 Digitally Controlled Oscillator – ADPLL: Small, Fast Frequency Changes

The Si544 can make small, fast frequency adjustments over a range of +/- 950 ppm (parts-per-million) around the device output frequency (set as described in previous sections). This mode is typically used in applications requiring a digitally controlled oscillator (DCO) for digital PLL or other types of frequency control loops. We refer to this type of application as an all-digital PLL or ADPLL.

For ADPLL applications and superior ADPLL performance, Silicon Labs recommends the Si548 I2C oscillator. The Si548 is a 6-pin device that features lower jitter than the Si544 while I2C lines are active, due to the different physical location of the SDA/SCL pins on the package.

The ADPLL mode uses a single 24 bit register, ADPLL_DELTA_M[23:0], to add an offset to the VCO frequency to affect the small frequency change. This offset is added in a synchronous fashion to prevent frequency discontinuities and can be updated as fast as the max I2C bus speed of 1 MHz will allow. The frequency offset can be positive or negative over a range of -950 ppm to +950 ppm with 0.0001164 ppm resolution.

The equation for this frequency change is simply,

$$\text{ADPLL_DELTA_M}[23:0] = \Delta \text{FoutPPM} / 0.0001164$$

Where $\Delta \text{FoutPPM}$ is the desired ppm change in output frequency, ADPLL_DELTA_M[23:0] is a two's complement 24 bit value, and 0.0001164 is a constant per-bit ppm value. The 24 bit ADPLL_DELTA_M[23:0] value is written into three sequential 8 bit registers in LSByte to MSByte order via I2C. Upon writing the MSByte, the frequency change takes effect. Below is an example VB to implement this feature. (Note that writing ADPLL_DELTA_M[23:0] = 0x000 will result in no frequency offset and return to the nominal output frequency.)

VB Code example for ADPLL (small frequency change) calculation and operation:

```
nAddr = Device I2C address
PPM_Delta = desired PPM frequency shift

Function Set_ADPLL(ByVal nAddr As UInteger, ByVal PPM_Delta As Double) As Integer
    Dim ADPLL_PPM_StepSize As Double = 0.0001164
    Dim ADPLL_Delta_M As Integer
    Dim Reg231 As UInteger = 0
    Dim Reg232 As UInteger = 0
    Dim Reg233 As UInteger = 0
    Dim ReturnCode As Integer = 0 '1=OK, -1 PPM requested is out of bounds
    If (PPM_Delta <= 950 And PPM_Delta >= -950) Then
        ADPLL_Delta_M = (PPM_Delta / ADPLL_PPM_StepSize)
        Reg231 = (ADPLL_Delta_M And &HFF)
        Reg232 = (ADPLL_Delta_M >> 8) And &HFF
        Reg233 = (ADPLL_Delta_M >> 16) And &HFF
        I2C_Write(nAddr, 0, 231, Reg231) 'write "Reg231" value to register 231 at nAddr, page 0 (LSByte)
        I2C_Write(nAddr, 0, 232, Reg232) 'write "Reg232" value to register 232 at nAddr, page 0
        I2C_Write(nAddr, 0, 233, Reg233) 'write "Reg233" value to register 233 at nAddr, page 0
    (MSByte)
        ReturnCode = 1
    Else
        ReturnCode = -1
    End If
    Return (ReturnCode)
End Function
```

5.7 Register Map Reference

Table 5.7. Register Map Reference Summary

Register (decimal)	Register Bit								Type	Reset Value
	7	6	5	4	3	2	1	0		
0	DEVICE_TYPE[7:0]								R	0x2C
7	RESET	<Reserved> = 3'b000			MS_ICAL 2	<Reserved> = 3'b000			R/W	0x00
17	<Unused>							ODC_OE	R/W	0x01
23	HSDIV[7:0]								R/W	0x54
24	<Unused>	LSDIV[2:0]			<Unused>	HSDIV[10:8]			R/W	0x00
26	FBDIV[7:0]								R/W	0x00
27	FBDIV[15:8]								R/W	0x00
28	FBDIV[23:16]								R/W	0x00
29	FBDIV[31:24]								R/W	0x00
30	FBDIV[39:32]								R/W	0x64
31	<Unused>					FBDIV[42:40]			R/W	0x00
69	FCAL_OV R	<Reserved> = 7'b0000001							R/W	0x01
231	ADPLL_DELTA_M[7:0]								R/W	0x00
232	ADPLL_DELTA_M[15:8]								R/W	0x00
233	ADPLL_DELTA_M[23:16]								R/W	0x00
255	<Reserved> = 6'b000000					PAGE[1:0]			R/W	0x00

Table 5.8. Register Bit Field Summary

Register Bit Field Name	Bit Field (#bits)	Register	Description
DEVICE_TYPE[7:0]	8	0	Read only value of 44 (dec) which represents the device type (Si544).
RESET	1	7	Set to 1 to reset device. Self clearing.
MS_ICAL2	1	7	Set to 1 to initiate FCAL. Self clearing.
ODC_OE	1	17	Set to 0 to disable the output clock, and remove output enable/disable control from the external OE or OEB pin. Set to 1 to return enable/disable control back to the OE or OEB pin.
HSDIV[10:0]	11	23-24	HSDIV is High-speed output divider value in unsigned 11-bit binary format. Valid divide values are from 5 to 2046, with values of 5-33 even or odd, and values 34-2046 restricted to even values only.

Register Bit Field Name	Bit Field (#bits)	Register	Description
LSDIV[2:0]	3	24	LSDIV sets a power-of-2 output divider. Values of 0,1,2,3,4,5,6,7 result in divide ratio of 1,2,4,8,16,32,32,32 respectively. Note that a value of 0 (divide-by-1) essentially bypasses this divider.
FBDIV[42:0]	43	26-31	The main DSPLL system feedback divide (FBDIV) value for Si54x. This 43 bit value is composed of an unsigned 11-bit integer value (FBDIV[42:32]) concatenated with a 32-bit fractional value (FBDIV[31:0]), for an 11.32 fixed point binary format. The valid range of the 11-bit integer part is from 60 to 2045
FCAL_OVR	1	69	FCAL Override: If set to 1, FCAL is bypassed. Clear to 0 to allow FCAL.
ADPLL_DELTA_M[23:0]	24	231-233	Digital word to effect small frequency shifts to base frequency. Value is 24 bit 2's complement causing a 0.0001164 ppm per bit shift in frequency. Positive values = positive freq shift, negative values = negative freq shift. Valid range is -8161513 to +8161512, representing a max PPM shift range of -950 ppm to +950 ppm, with 0 value representing 0 PPM shift. Writing a new ADPLL_DELTA_M value will take effect upon writing to the MSByte (Register 233). Therefore, value updates should follow the sequence of writing in register order Reg 231...Reg 232...Reg 233.
PAGE[1:0]	2	255	Sets which page of registers the I2C port is reading/writing. The size of a page is 256 bytes which is the addressable range of an I2C " set address " command. The value of PAGE is multiplied by 256 and added to what " set address " has set. Physically, the 2 PAGE bits become bits [9:8] of the device's internal register map address. This mechanism allows for more than 256 registers to be addressed within the 8 bit I2C " set address " limitation.

5.8 Si544 Frequency Planner VB Code

```

Module Main
'
' Si54x Frequency Planner Code
'
'
' Set Target device type, Speed grade, and desired output frequency
'
Public Device As Integer = 544      ' 544 or 549 only
Public SpeedGrade As String = "C"  ' Can only be "A" or "B" or "C"
Public Output_Freq As Double = 312500000.0 ' Output frequency in Hz (initially set to 312.5 MHz)

' Set in 'SetLimits" function...
Public Fvco_max As Double ' Fvco Max per Table 5.3
Public Fvco_min As Double ' Fvco Min per Table 5.3
Public Xtal_freq As Double ' Xtal_Freq per Table 5.3
Public Fout_min As Double ' Minimum output frequency
Public Fout_max As Double ' Maximum output frequency

Sub Main()
'
' Device divider limits (see Tables 5.1 & 5.2)
'
Dim HSDIV_UpperLimit As Integer = 2046
Dim HSDIV_LowerLimit As Integer = 5
Dim HSDIV_LowerLimit_Odd As Integer = 5      ' min count for odd HSDIV divisor
Dim HSDIV_UpperLimit_Odd As Integer = 33     ' max count for odd HSDIV divisor
Dim LSDIV_UpperLimit As Integer = 5
Dim LSDIV_LowerLimit As Integer = 0
Dim FBDIV_UpperLimit As Double = 2045 + ((2 ^ 32 - 1) / (2 ^ 32))
Dim FBDIV_LowerLimit As Double = 60.0
'
' Working variables
'
Dim Min_HSLS_Div As Double
Dim LSDIV_Div As Double      ' actual LSDIV divide ratio
Dim LSDIV_Reg As Integer    ' LSDIV as encoded in power of 2 for device register use
Dim HSDIV As Double
Dim FBDIV As Double
Dim Fvco As Double
Dim FBDIV_Int As UInteger
Dim FBDIV_Frac As UInteger
Dim Reg23 As UInteger = 0   ' HSDIV[7:0]
Dim Reg24 As UInteger = 0   ' OD_LSDIV[2:0], HSDIV[10:8] (*2^4, /2^8)
Dim Reg26 As UInteger = 0   ' FBDIV[7:0]
Dim Reg27 As UInteger = 0   ' FBDIV[15:8] (/2^8)
Dim Reg28 As UInteger = 0   ' FBDIV[23:16] (/2^16)
Dim Reg29 As UInteger = 0   ' FBDIV[31:24] (/2^24)
Dim Reg30 As UInteger = 0   ' FBDIV[39:32] (/2^32)
Dim Reg31 As UInteger = 0   ' FBDIV[42:40] (/2^40)
'
' Set device limits based on device type and speed grade.
' (Checks if desired output frequency is valid based on device and speed grade)
'
If SetLimits(Device, SpeedGrade, Output_Freq) = 0 Then
'
' If limits are set and output frequency is valid, calculate frequency plan...
'
' *****
' Step 1: Find theoretical HSDIV *LSDIV value based on lowest valid VCO frequency...
' (Assumes "Output_Freq" has been tested and is in valid range for the device grade
according to Table 5.3)
'
Min_HSLS_Div = Fvco_min / Output_Freq      ' Floating point HS*LS div value. Remember to first
bounds check Output_Freq!

' Step 2: Find LSDIV divisor value given Min_HSLS_Div value
'
LSDIV_Div = Math.Ceiling(Min_HSLS_Div / HSDIV_UpperLimit) ' Divisor value of LSDIV, NOT yet
encoded as power of 2

```

```

    If (LSDIV_Div > 32) Then LSDIV_Div = 32 ' clip at 32 (max LSDIV divisor)
    '
    'Encode LSDIV divisor value into next nearest 'power of 2' value if not already. This will be
LSDIV_Reg
    '
    LSDIV_Reg = Math.Ceiling(Math.Log(LSDIV_Div, 2)) ' LSDIV_Reg now encoded as proper power of
2. Will range from 0 to 5.

    ' Adjust LSDIV_Div (holder of divisor) based on rounded power of 2 value in LSDIV_Reg
LSDIV_Div = 2 ^ LSDIV_Reg 'LSDIV_Div divisor now synchronized to actual LSDIV_Reg.
    '
    'Step 3: Find HSDIV divisor value using known LSDIV divisor
    '
HSDIV = Math.Ceiling(Min_HSLS_Div / LSDIV_Div)
If ((HSDIV >= HSDIV_LowerLimit_Odd) And (HSDIV <= HSDIV_UpperLimit_Odd)) Then
    HSDIV = HSDIV ' Leaves HSDIV as even or odd only if HSDIV is from 5 to 33.
Else
    If ((HSDIV Mod 2) <> 0) Then 'If HSDIV is an odd value...
        HSDIV = HSDIV + 1 '...make it even by rounding up
    End If 'If already even, leave it alone
End If
    '
    ' Step 4: Now calculate Fvco and FBDIV
    '
Fvco = (HSDIV * LSDIV_Div * Output_Freq) 'Calculate Fvco based on valid HSDIV,LSDIV, and Fout
FBDIV = Fvco / Xtal_freq 'Finally, calculate FBDIV based on xtal freq

'Calculate 11.32 fixed point FBDIV value (MCTL_M)
'Extract Integer part
FBDIV_Int = Int(FBDIV)
'Extract fractional part
FBDIV = (FBDIV - FBDIV_Int)
FBDIV = FBDIV * (2 ^ 32)
FBDIV_Frac = Int(FBDIV)
'
'Generate Register values based on LSDIV, HSDIV, and FBDIV (MCTL_M)
'
Reg23 = (HSDIV And &HFF)
Reg24 = ((HSDIV >> 8) And &H7) Or ((LSDIV_Reg And &H7) << 4)

Reg26 = (FBDIV_Frac And &HFF)
Reg27 = (FBDIV_Frac >> 8) And &HFF
Reg28 = (FBDIV_Frac >> 16) And &HFF
Reg29 = (FBDIV_Frac >> 24) And &HFF
Reg30 = (FBDIV_Int) And &HFF
Reg31 = (FBDIV_Int >> 8) And &H7

    '*****
Else
    Console.WriteLine("*** Device invalid or Device limits exceeded. Frequency plan not calculated.")
End If

End Sub
'
' Sets device limits according to Table 5.3
' Returns 0 if limits are set and output frequency is valid
' Returns -1 if limits not found or output frequency is invalid
Function SetLimits(ByVal Device As Integer, ByVal SpeedGrade As String, ByVal Output_Freq As Double) As
Integer
    Dim ReturnCode As Integer
    ReturnCode = 0
    If Device = 544 Then
        Xtal_freq = 55050000.0
        If SpeedGrade = "A" Then
            Fvco_min = 10800000000.0
            Fvco_max = 12550082103.0
            Fout_min = 200000.0
            Fout_max = 1500000000.0
            If ((Output_Freq < Fout_min) Or (Output_Freq > Fout_max)) Then
                ReturnCode = -1
            End If
        ElseIf SpeedGrade = "B" Then
            Fvco_min = 10800000000.0

```

```
Fvco_max = 12109728345.0
Fout_min = 200000.0
Fout_max = 800000000.0
If ((Output_Freq < Fout_min) Or (Output_Freq > Fout_max)) Then
    ReturnCode = -1
End If
ElseIf SpeedGrade = "C" Then
    Fvco_min = 10800000000.0
    Fvco_max = 12109728345.0
    Fout_min = 200000.0
    Fout_max = 325000000.0
    If ((Output_Freq < Fout_min) Or (Output_Freq > Fout_max)) Then
        ReturnCode = -1
    End If
Else
    ReturnCode = -1
End If
ElseIf Device = 549 Then
    Xtal_freq = 152600000.0
    If SpeedGrade = "A" Then
        Fvco_min = 10800000000.0
        Fvco_max = 12511886114.0
        Fout_min = 200000.0
        Fout_max = 1500000000.0
        If ((Output_Freq < Fout_min) Or (Output_Freq > Fout_max)) Then
            ReturnCode = -1
        End If
    ElseIf SpeedGrade = "B" Then
        Fvco_min = 10800000000.0
        Fvco_max = 12206718160.0
        Fout_min = 200000.0
        Fout_max = 800000000.0
        If ((Output_Freq < Fout_min) Or (Output_Freq > Fout_max)) Then
            ReturnCode = -1
        End If
    ElseIf SpeedGrade = "C" Then
        Fvco_min = 10800000000.0
        Fvco_max = 12206718160.0
        Fout_min = 200000.0
        Fout_max = 325000000.0
        If ((Output_Freq < Fout_min) Or (Output_Freq > Fout_max)) Then
            ReturnCode = -1
        End If
    Else
        ReturnCode = -1
    End If
Else
    ReturnCode = -1
End If
Return (ReturnCode)
End Function
End Module
-----
```

5.9 Table of Common Frequencies for Si544 (55.05 MHz xtal)

Fout (MHz)	LSDIV	HSDIV	FBDIV	Fvco (GHz)	Reg 23	Reg 24	Reg 26	Reg 27	Reg 28	Reg 29	Reg 30	Reg 31
70.656	0	154	197.6571117	10.881024	9Ah	00h	34h	79h	38h	A8h	C5h	00h
100	0	108	196.1852861	10.8	6Ch	00h	FBh	E8h	6Eh	2Fh	C4h	00h
122.88	0	88	196.4294278	10.81344	58h	00h	D6h	FAh	EEh	6Dh	C4h	00h
125	0	88	199.818347	11	58h	00h	AAh	2Fh	7Fh	D1h	C7h	00h
148.351648	0	74	199.419109	10.97802195	4Ah	00h	B1h	BAh	4Ah	6Bh	C7h	00h
148.5	0	74	199.6185286	10.989	4Ah	00h	19h	E4h	57h	9Eh	C7h	00h
148.945454	0	74	200.2173224	11.0219636	4Ah	00h	33h	70h	A2h	37h	C8h	00h
150	0	72	196.1852861	10.8	48h	00h	FBh	E8h	6Eh	2Fh	C4h	00h
153.6	0	72	200.893733	11.0592	48h	00h	15h	AFh	CBh	E4h	C8h	00h
155.52	0	70	197.7547684	10.8864	46h	00h	59h	80h	38h	C1h	C5h	00h
156.25	0	70	198.6830154	10.9375	46h	00h	93h	19h	DAh	A Eh	C6h	00h
168.04	0	66	201.4648501	11.09064	42h	00h	24h	6Bh	00h	77h	C9h	00h
168.75	0	64	196.1852861	10.8	40h	00h	FBh	E8h	6Eh	2Fh	C4h	00h
200	0	54	196.1852861	10.8	36h	00h	FBh	E8h	6Eh	2Fh	C4h	00h
212.5	0	52	200.7266122	11.05	34h	00h	56h	41h	03h	BAh	C8h	00h
245.76	0	44	196.4294278	10.81344	2Ch	00h	D6h	FAh	EEh	6Dh	C4h	00h
250	0	44	199.818347	11	2Ch	00h	AAh	2Fh	7Fh	D1h	C7h	00h
270	0	40	196.1852861	10.8	28h	00h	FBh	E8h	6Eh	2Fh	C4h	00h
311.04	0	36	203.4049046	11.19744	24h	00h	79h	D4h	A7h	67h	CBh	00h
312.5	0	36	204.359673	11.25	24h	00h	05h	88h	13h	5Ch	CCh	00h
322.265625	0	34	199.0378065	10.95703125	22h	00h	7Ah	B0h	ADh	09h	C7h	00h
400	0	27	196.1852861	10.8	1Bh	00h	FBh	E8h	6Eh	2Fh	C4h	00h
425	0	26	200.7266122	11.05	1Ah	00h	56h	41h	03h	BAh	C8h	00h
491.52	0	22	196.4294278	10.81344	16h	00h	D6h	FAh	EEh	6Dh	C4h	00h
500	0	22	199.818347	11	16h	00h	AAh	2Fh	7Fh	D1h	C7h	00h
614.4	0	18	200.893733	11.0592	12h	00h	15h	AFh	CBh	E4h	C8h	00h
622.08	0	18	203.4049046	11.19744	12h	00h	79h	D4h	A7h	67h	CBh	00h
644.53125	0	17	199.0378065	10.95703125	11h	00h	7Ah	B0h	ADh	09h	C7h	00h
750	0	15	204.359673	11.25	0Fh	00h	05h	88h	13h	5Ch	CCh	00h
800	0	14	203.4514078	11.2	0Eh	00h	59h	76h	8Fh	73h	CBh	00h

5.10 I2C Interface

Configuration and operation of the Si544 is controlled by reading and writing to the RAM space using the I2C interface. The device operates in slave mode with 7-bit addressing and can operate in Standard-Mode (100 kbps), Fast-Mode (400 kbps), or Fast-Mode Plus (1 Mbps). Burst data transfer with auto address increments are also supported.

The I2C bus consists of a bidirectional serial data line (SDA) and a serial clock input (SCL). Both the SDA and SCL pins must be connected to the VDD supply via an external pull-up as recommended by the I2C specification. The Si544 7-bit I2C slave address is user-customized during the part number configuration process.

Data is transferred MSB first in 8-bit words as specified by the I2C specification. A write command consists of a 7-bit device (slave) address + a write bit, an 8-bit register address, and 8 bits of data as shown in the figure below.

A write burst operation is also shown where every additional data word is written using an auto-incremented address.

Write Operation – Single Byte



Write Operation - Burst (Auto Address Increment)

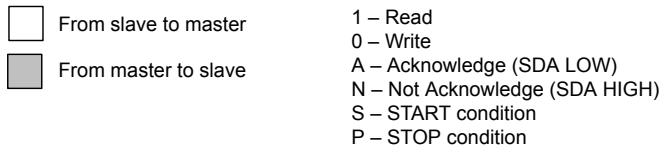
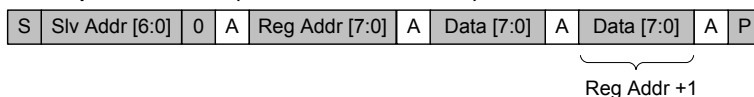
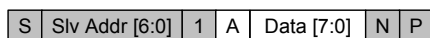
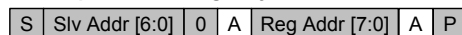


Figure 5.3. I2C Write Operation

A read operation is performed in two stages. A data write is used to set the register address, then a data read is performed to retrieve the data from the set address. A read burst operation is also supported. This is shown in the figure below.

Read Operation – Single Byte



Read Operation - Burst (Auto Address Increment)

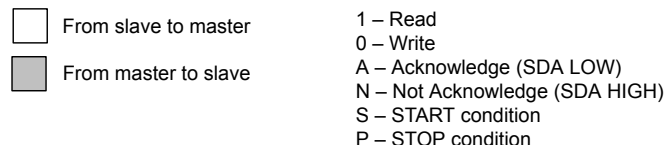
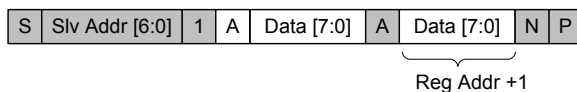


Figure 5.4. I2C Read Operation

The timing specifications and timing diagram for the I2C bus is compatible with the I2C-Bus standard. SDA timeout is supported for compatibility with SMBus interfaces.

The I2C bus can be operated at a bus voltage of 1.71 to 3.63 V and should be the same voltage as the Si544 VDD.

6. Package Outline

6.1 Package Outline (5x7 mm)

The figure below illustrates the package details for the 5x7 mm Si544. The table below lists the values for the dimensions shown in the illustration.

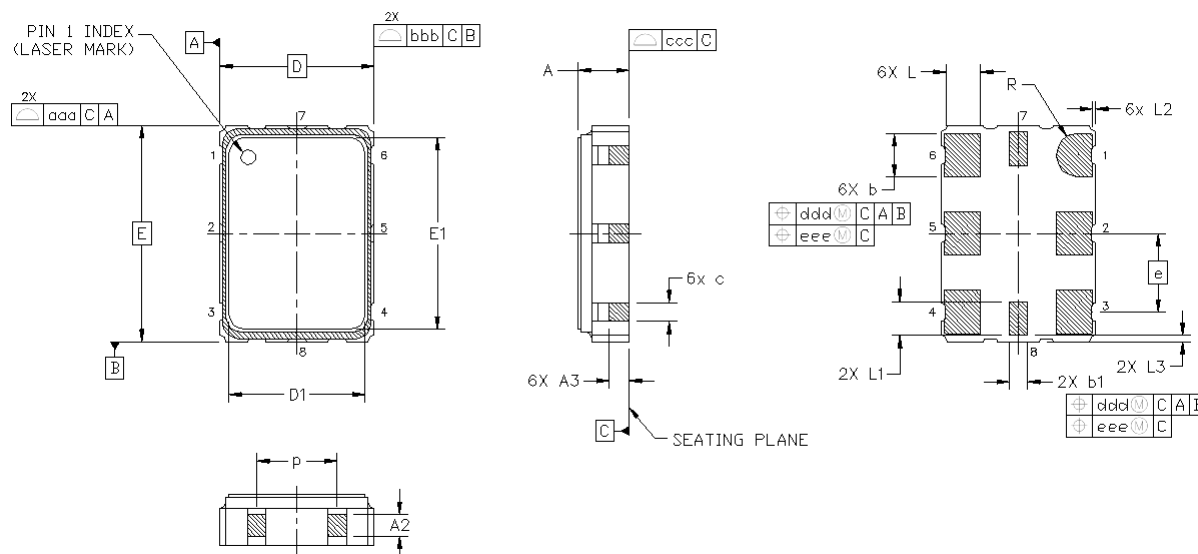


Figure 6.1. Si544 (5x7 mm) Outline Diagram

Table 6.1. Package Diagram Dimensions (mm)

Dimension	Min	Nom	Max	Dimension	Min	Nom	Max
A	1.07	1.18	1.33	L	1.07	1.17	1.27
A2	0.40	0.50	0.60	L1	1.00	1.10	1.20
A3	0.45	0.55	0.65	L2	0.05	0.10	0.15
b	1.30	1.40	1.50	L3	0.15	0.20	0.25
b1	0.50	0.60	0.70	p	1.70	--	1.90
c	0.50	0.60	0.70	R	0.70 REF		
D	5.00 BSC			aaa	0.15		
D1	4.30	4.40	4.50	bbb	0.15		
e	2.54 BSC			ccc	0.08		
E	7.00 BSC			ddd	0.10		
E1	6.10	6.20	6.30	eee	0.05		

Notes:

1. All dimensions shown are in millimeters (mm) unless otherwise noted.
2. Dimensioning and Tolerancing per ANSI Y14.5M-1994.

6.2 Package Outline (3.2x5 mm)

The figure below illustrates the package details for the 3.2x5 mm Si544. The table below lists the values for the dimensions shown in the illustration.

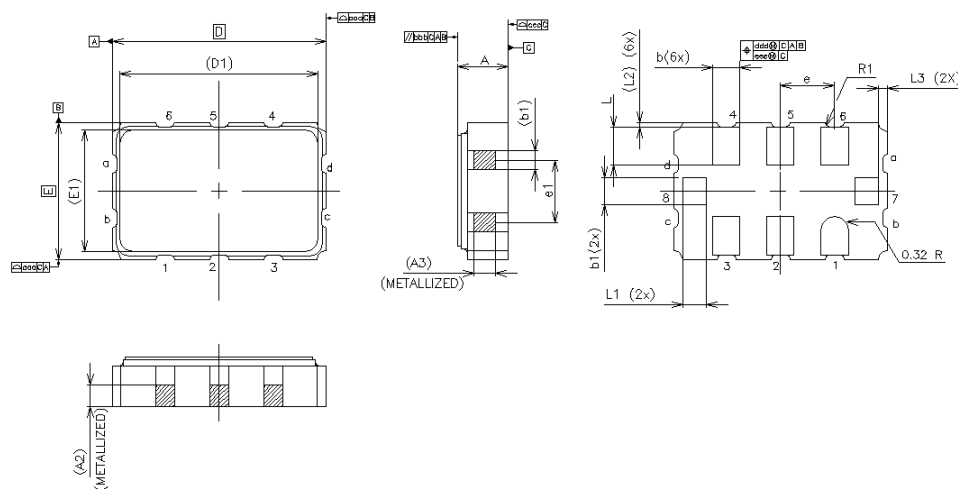


Figure 6.2. Si544 (3.2x5 mm) Outline Diagram

Table 6.2. Package Diagram Dimensions (mm)

Dimension	MIN	NOM	MAX	Dimension	MIN	NOM	MAX
A	1.02	1.17	1.33	E1	2.85 BSC		
A2	0.50	0.55	0.60	L	0.8	0.9	1.0
A3	0.45	0.50	0.55	L1	0.45	0.55	0.65
b	0.54	0.64	0.74	L2	0.05	0.10	0.15
b1	0.54	0.64	0.75	L3	0.15	0.20	0.25
D	5.00 BSC			aaa	0.15		
D1	4.65 BSC			bbb	0.15		
e	1.27 BSC			ccc	0.08		
e1	1.625 TYP			ddd	0.10		
E	3.20 BSC			eee	0.05		

Notes:

1. All dimensions shown are in millimeters (mm) unless otherwise noted.
2. Dimensioning and Tolerancing per ANSI Y14.5M-1994.

6.3 Package Outline (2.5x3.2 mm)

The figure below illustrates the package details for the 2.5x3.2 mm Si544. The table below lists the values for the dimensions shown in the illustration.

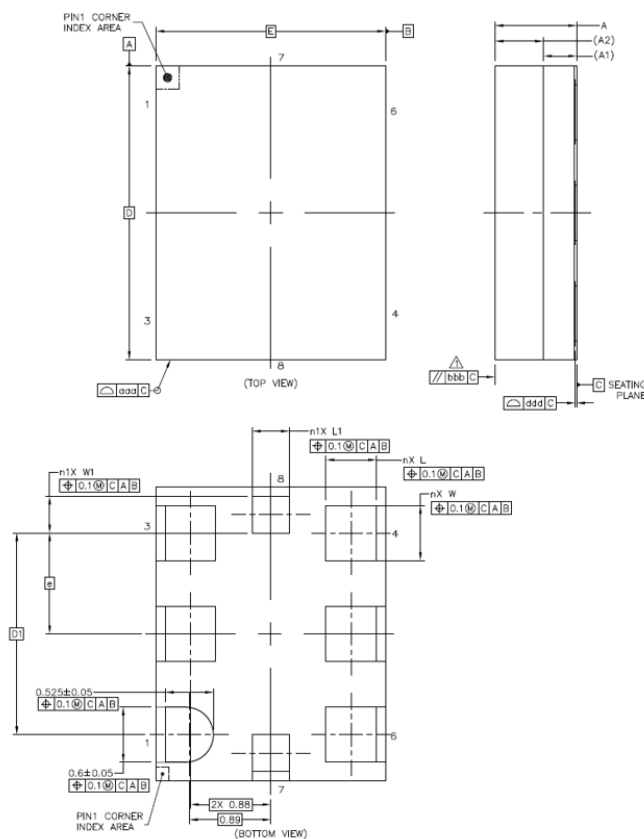


Figure 6.3. Si544 (2.5x3.2 mm) Outline Diagram

Table 6.3. Package Diagram Dimensions (mm)

Dimension	MIN	NOM	MAX	Dimension	MIN	NOM	MAX
A	—	—	1	L1	0.35	0.4	0.45
A1	0.36 REF			e	1.1 BSC		
A2	0.53 REF			n	5		
D	3.2 BSC			n1	2		
E	2.5 BSC			D1	2.2 BSC		
W	0.55	0.6	0.65	aaa	0.10		
L	0.5	0.55	0.6	bbb	0.10		
W1	0.35	0.4	0.45	ddd	0.08		

Notes:

1. The dimensions in parentheses are reference.
2. All dimensions shown are in millimeters (mm) unless otherwise noted.
3. Dimensioning and Tolerancing per ANSI Y14.5M-1994.
4. Recommended card reflow profile is per the JEDEC/IPC J-STD-020 specification for Small Body Components.

7. PCB Land Pattern

7.1 PCB Land Pattern (5x7 mm)

The figure below illustrates the 5x7 mm PCB land pattern for the Si544. The table below lists the values for the dimensions shown in the illustration.

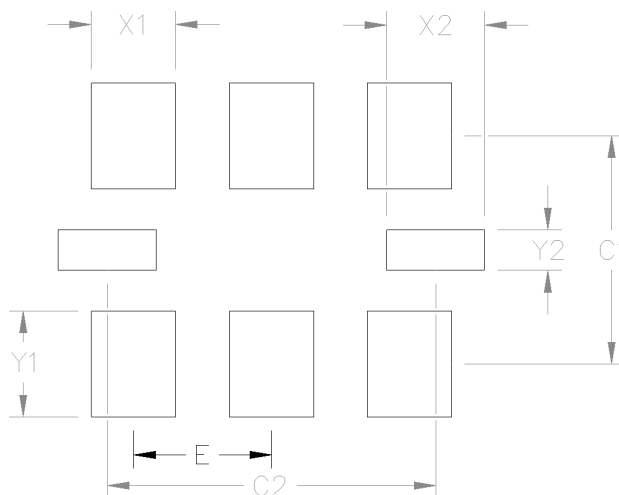


Figure 7.1. Si544 (5x7 mm) PCB Land Pattern

Table 7.1. PCB Land Pattern Dimensions (mm)

Dimension	(mm)	Dimension	(mm)
C1	4.20	Y1	1.95
C2	6.05	X2	1.80
E	2.54	Y2	0.75
X1	1.55		

Notes:

General

1. All dimensions shown are in millimeters (mm) unless otherwise noted.
2. Dimensioning and Tolerancing is per the ANSI Y14.5M-1994 specification.
3. This Land Pattern Design is based on the IPC-7351 guidelines.
4. All dimensions shown are at Maximum Material Condition (MMC). Least Material Condition (LMC) is calculated based on a Fabrication Allowance of 0.05 mm.

Solder Mask Design

1. All metal pads are to be non-solder mask defined (NSMD). Clearance between the solder mask and the metal pad is to be 60 μm minimum, all the way around the pad.

Stencil Design

1. A stainless steel, laser-cut and electro-polished stencil with trapezoidal walls should be used to assure good solder paste release.
2. The stencil thickness should be 0.125 mm (5 mils).
3. The ratio of stencil aperture to land pad size should be 1:1.

Card Assembly

1. A No-Clean, Type-3 solder paste is recommended.
2. The recommended card reflow profile is per the JEDEC/IPC J-STD-020 specification for Small Body Components.

7.2 PCB Land Pattern (3.2x5 mm)

The figure below illustrates the 3.2x5.0 mm PCB land pattern for the Si544. The table below lists the values for the dimensions shown in the illustration.

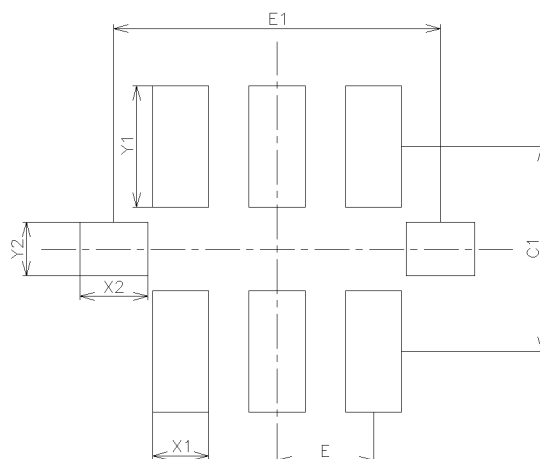


Figure 7.2. Si544 (3.2x5 mm) PCB Land Pattern

Table 7.2. PCB Land Pattern Dimensions (mm)

Dimension	(mm)	Dimension	(mm)
C1	2.70	X2	0.90
E	1.27	Y1	1.60
E1	4.30	Y2	0.70
X1	0.74		

Notes:

General

1. All dimensions shown are in millimeters (mm) unless otherwise noted.
2. Dimensioning and Tolerancing is per the ANSI Y14.5M-1994 specification.
3. This Land Pattern Design is based on the IPC-7351 guidelines.
4. All dimensions shown are at Maximum Material Condition (MMC). Least Material Condition (LMC) is calculated based on a Fabrication Allowance of 0.05 mm.

Solder Mask Design

1. All metal pads are to be non-solder mask defined (NSMD). Clearance between the solder mask and the metal pad is to be 60 μm minimum, all the way around the pad.

Stencil Design

1. A stainless steel, laser-cut and electro-polished stencil with trapezoidal walls should be used to assure good solder paste release.
2. The stencil thickness should be 0.125 mm (5 mils).
3. The ratio of stencil aperture to land pad size should be 1:1.

Card Assembly

1. A No-Clean, Type-3 solder paste is recommended.
2. The recommended card reflow profile is per the JEDEC/IPC J-STD-020 specification for Small Body Components.

7.3 PCB Land Pattern (2.5x3.2 mm)

The figure below illustrates the 2.5x3.2 mm PCB land pattern for the Si544. The table below lists the values for the dimensions shown in the illustration.

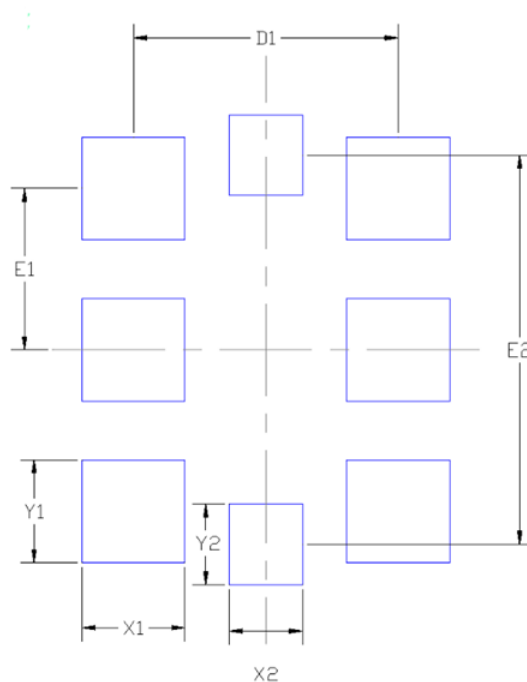


Figure 7.3. Si544 (2.5x3.2 mm) PCB Land Pattern

Table 7.3. PCB Land Pattern Dimensions (mm)

Dimension	Description	Value (mm)
X1	Width - leads on long sides	0.7
Y1	Height - leads on long sides	0.7
X2	Width - single leads on short sides	0.5
Y2	Height - single leads on short sides	0.55
D1	Pitch in X directions of XL, Y1 leads	1.80
E1	Lead pitch X1, Y1 leads	1.10
E2	Lead pitch X2, Y2 leads	2.65

Dimension	Description	Value (mm)
<p>Notes:</p> <p>General</p> <ol style="list-style-type: none">1. All dimensions shown are in millimeters (mm) unless otherwise noted.2. Dimensioning and Tolerancing is per the ANSI Y14.5M-1994 specification.3. This Land Pattern Design is based on the IPC-7351 guidelines.4. All dimensions shown are at Maximum Material Condition (MMC). Least Material Condition (LMC) is calculated based on a Fabrication Allowance of 0.05 mm. <p>Solder Mask Design</p> <ol style="list-style-type: none">1. All metal pads are to be non-solder mask defined (NSMD). Clearance between the solder mask and the metal pad is to be 60 μm minimum, all the way around the pad. <p>Stencil Design</p> <ol style="list-style-type: none">1. A stainless steel, laser-cut and electro-polished stencil with trapezoidal walls should be used to assure good solder paste release.2. The stencil thickness should be 0.125 mm (5 mils).3. The ratio of stencil aperture to land pad size should be 0.8:1 for the pads. <p>Card Assembly</p> <ol style="list-style-type: none">1. A No-Clean, Type-3 solder paste is recommended.2. The recommended card reflow profile is per the JEDEC/IPC J-STD-020 specification for Small Body Components.		

8. Top Marking (5x7 and 3.2x5 Packages)

The figure below illustrates the mark specification for the Si544 5x7 and 3.2x5 package sizes. The table below lists the line information.

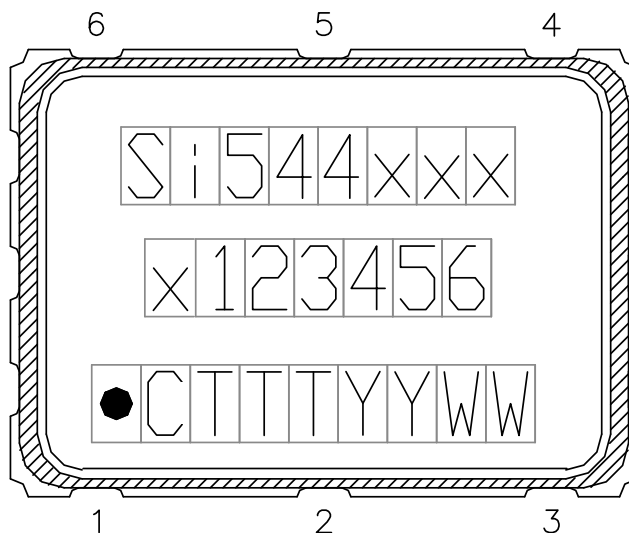


Figure 8.1. Mark Specification

Table 8.1. Si544 Top Mark Description

Line	Position	Description
1	1–8	"Si544", xxx = Ordering Option 1, Option 2, Option 3 (e.g. Si544AAA)
2	1	x = Frequency Range Supported as described in the Ordering Guide
	2–7	6-digit custom Frequency Code as described in the Ordering Guide
3	Trace Code	
	Position 1	Pin 1 orientation mark (dot)
	Position 2	Product Revision (C)
	Position 3–5	Tiny Trace Code (3 alphanumeric characters per assembly release instructions)
	Position 6–7	Year (last two digits of the year), to be assigned by assembly site (ex: 2017 = 17)
	Position 8–9	Calendar Work Week number (1–53), to be assigned by assembly site

9. Top Marking (2.5x3.2 Package)

The figure below illustrates the mark specification for the Si544 2.5x3.2 package size. The table below lists the line information.

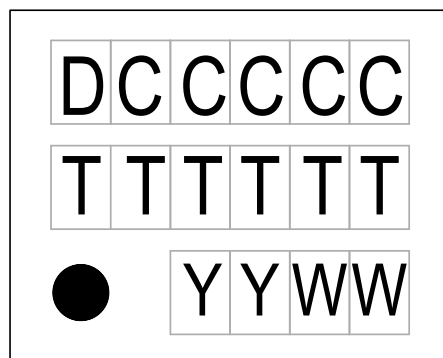


Figure 9.1. Mark Specification

Table 9.1. Si544 Top Mark Description

Line	Position	Description
1	1–6	D = Si544, CCCCC = Custom Mark Code
2	Trace Code	
	1–6	Six digit trace code per assembly release instructions
3	Position 1	Pin 1 orientation mark (dot)
	Position 2–3	Year (last two digits of the year), to be assigned by assembly site (ex: 2017 = 17)
	Position 4–5	Calendar Work Week number (1–53), to be assigned by assembly site

10. Revision History

Revision 1.3

June 2021

- Updated Ordering Guide and topmark for RevC silicon
- Added HCSL-Fast (faster tR/tF) ordering option
- Updated Table 2.1, Powerup VDD Ramp Rate
- Updated Table 2.2, Settling Time for Large Frequency Change

Revision 1.2

September 2020

- Updated Table 2.1, Powerup VDD Ramp Rate and LVDS Swing

Revision 1.1

November 2019

- Added 2.5x3.2 mm package option.

Revision 1.0

July, 2018

- Added 20 ppm total stability option.

Revision 0.75

March, 2018

- Added 25 ppm total stability option.

Revision 0.5

October 27, 2017

- Initial release.

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